

Final-December 11, 1998  
EE524-Dr. Dickerson  
**Answer 4 out of 5 problems**

Name:

Code Name for posting grade on web:

Social Security Number

Problem Number	Points	Score
1	25	
2	25	
3	25	
4	25	
5	25	

Name:

1. Linear Phase

A type 3 linear phase filter has an antisymmetric impulse response and an odd length. The condition on the impulse response coefficients is:

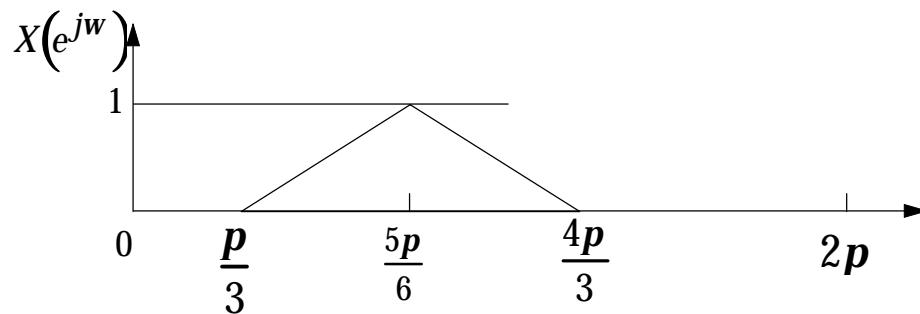
$$h[n] = -h[N - n] \quad 0 \leq n \leq N$$

- a) For  $N = 6$  show that a filter that meets this symmetry condition must be linear phase. Calculate the group delay for this filter. (10 points)
- b) Prove that the zeros for this anti-symmetric linear phase filter must be in complex conjugate reciprocal pairs when the filter coefficients are real. (10 points)
- c) Can an all pass FIR filter be linear phase - why or why not? (5 points)

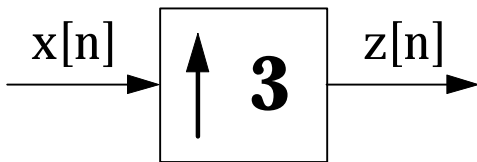
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## 2. Multirate Signal Processing

Consider a sequence  $x[n]$  with the spectrum shown below.



- Let  $y[n] = x[3n]$ . What is the spectrum of  $y[n]$  in terms of  $X(e^{j\omega})$ ? Sketch the spectrum. (10 points)
- Show how to recover  $x[n]$  from  $y[n]$  using filters and multirate building blocks. (10 points)
- Sketch the spectrum of  $z[n]$  in terms of  $X(e^{j\omega})$ . (5 points)



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3. DFTs

A signal consisting of a single sinusoid of frequency 2.0 KHz is sampled at 8 kHz.

a) 32 samples of the signal are taken and the 32 point DFT is calculated. Give a definition of resolution in the DFT. What is the resolution in the 32 point DFT. Sketch the magnitude of the DFT. (10%)

b) Now zero pad the sequence to length 128 points and take the 128 point DFT. What is the resolution now? Sketch the DFT magnitude. (5%)

c) Take the result of step (b) (the zero-padded DFT  $X_p(k)$ ) and form the sequence:

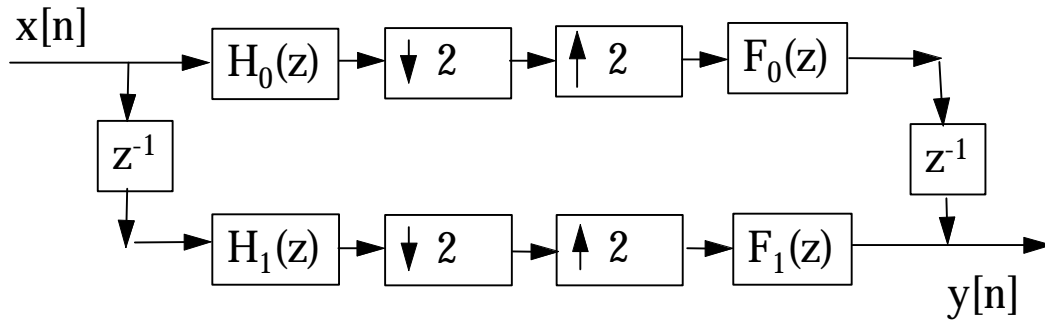
$$\bar{X}(k) = X_p(2k) \quad k = 0, 1, 2, \dots, 63$$

Take the 64 point inverse DFT. Sketch the result. (10%)

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4. Polyphase Filters

Consider the modified QMF bank shown below. Assume that all the filters are FIR.



- Express  $Y(z)$  in terms of  $X(z)$ . Write  $Y(z)$  in the form:  $Y(z) = T(z)X(z) + A(z)X(-z)$  (10 Points)
- Show that using the substitutions  $H_1(z) = H_0(-z)$  along with  $F_0(z) = H_0(z)$  and  $F_1(z) = H_1(z)$  will cancel out aliasing. Write the distortion transfer function  $T(z)$  in terms of  $H_0(z)$ . (7 Points)
- Let  $H_0(z)$  be a real coefficient linear phase FIR low pass filter of order  $N$ . Simplify  $T(z)$  and show that there is no phase distortion. Also show that  $N$  must be even to avoid  $T(e^{j\pi/2}) = 0$  (8 points)

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5. FIR filters using the Window method

Consider the following ideal frequency response for a multiband filter:

$$H_d(\omega) = \begin{cases} 0 & 0 \leq |\omega| < 0.3p \\ e^{-j\omega M/4} & 0.3p \leq |\omega| < 0.6p \\ 0.5 e^{-j\omega M/2} & 0.6p \leq |\omega| \leq p \end{cases}$$

- Sketch the phase and magnitude response of this filter. (5 points)
- Determine the ideal impulse response of this filter,  $h_d[n]$  (7 points)
- The impulse response  $h_d[n]$  is multiplied by a Blackman window of length 50 resulting in a linear phase FIR system. Determine the set of approximation error specifications there are satisfied by this FIR filter, (i.e. determine the parameters  $d_1, d_2, d_3, A, B, \omega_{p1}, \omega_s, \omega_{p2}$ ).

$$\left| H(e^{j\omega}) \right| \leq d_1 \quad 0 \leq \omega \leq \omega_s$$

$$A - d_2 \leq \left| H(e^{j\omega}) \right| \leq A + d_2 \quad \omega_{p1} \leq \omega \leq \omega_{p2}$$

$$B - d_3 \leq \left| H(e^{j\omega}) \right| \leq B + d_3 \quad \omega_{p2} \leq \omega \leq p$$